FPGA based High Speed Double Precision Floating Point Divider

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ABSTRACT

Floating point arithmetic is widely used in many areas, especially scientific computation and signal processing. For many signal processing, and graphics applications, it is acceptable to trade off some accuracy (in the least significant bit positions) for faster and better implementations. Division is the third basic operation of arithmetic. However, most of these modern applications need higher frequency or low latency of operations with minimal area occupancy. In this paper we describe an implementation of high speed IEEE 754 double precision floating point divider using digit recurrence algorithm and targeted for Xilinx Virtex-6 Field Programmable Gate Array. Verilog is used to implement the design. The implemented design achieves 344.89 MFlops and this design occupies 653 slices. It handles the overflow, underflow and rounding mode.

Keywords

Double precision, Floating point, Divider, FPGA, IEEE-754, and Virtex6

1. INTRODUCTION

Floating point numbers are one possible way of representing real numbers in binary format. The IEEE 754 standard presents two different floating point formats, Binary interchange format and Decimal interchange format. Multiplying floating point numbers is a critical requirement for DSP applications involving large dynamic range. This paper focuses on double precision floating point binary interchange format. Figure.1 shows the IEEE 754 double precision floating point binary format representation. It consists of a one bit sign (S), an eleven bits exponent (E), and a fifty two bits fraction (M or Mantissa). An extra bit is added to the fraction to form what is called the significand1. If the exponent is greater than 0 and smaller than 2047, and there is 1 in the MSB of the significand then the number is said to be a normalized number, Significand is the mantissa with an extra MSB bit.

$$Z = (-1^S) * 2^{-(E-Bias)} * (1.M)$$
 Where $M = m_{51} 2^{-1} + m_{50} 2^{-2} + m_{49} 2^{-3} + ... + m_1 2^{-51} + m_0 2^{-3}$ Bias = 1023.

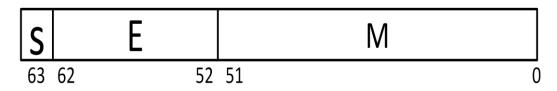


Figure 1: IEEE Double precision floating point format

2. IMPLEMENTATION OF HIGH SPEED DOUBLE PRECISION FLOATING POINT DIVIDER

.The double precision floating point divider performs operation such as division. The Black box view of double precision floating point divider is shown in figure 2.

The input signals to the top level module are

- 1. Clk
- 2. Rst
- 3. Enable
- 4. Rmode (Rounding mode)

Rounding Modes selected for various bit combinations

Bit combination	Rounding Mode
00	round_nearest_even
01	round_to_zero
10	round_up
11	round_down

- 5. Opa (64 bits)
- 6. Opb (64 bits)

The output signals from the module are

- 1. Out (output from operation, 64 bits)
- 2. Ready
- 3. Underflow
- 4. Overflow
- 5. Inexact
- 6. Exception
- 7. Invalid

The block diagram and inter-connections of the three submodules of the double precision floating point divider is shown in figure 3 and 4 respectively. The sub modules of double precision floating point divider are listed below.

- 1. Fp_div_int
- 2. Fp_round
- 3.Fp_exception

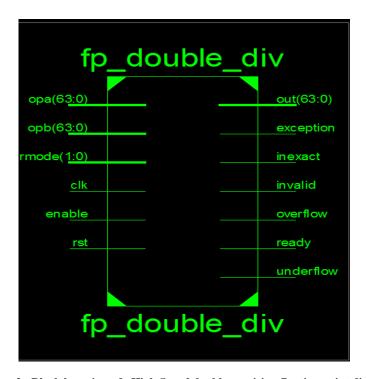


Figure 2: Black box view of High Speed double precision floating point divider

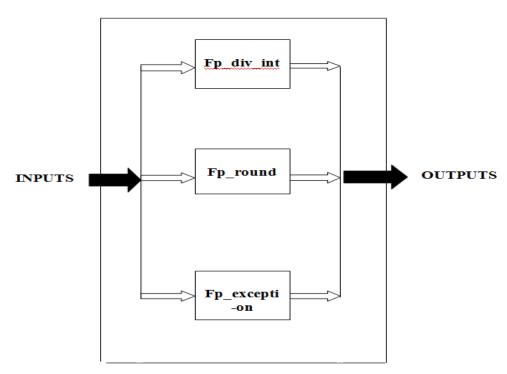


Figure 3: Block diagram of High Speed double precision floating point divider

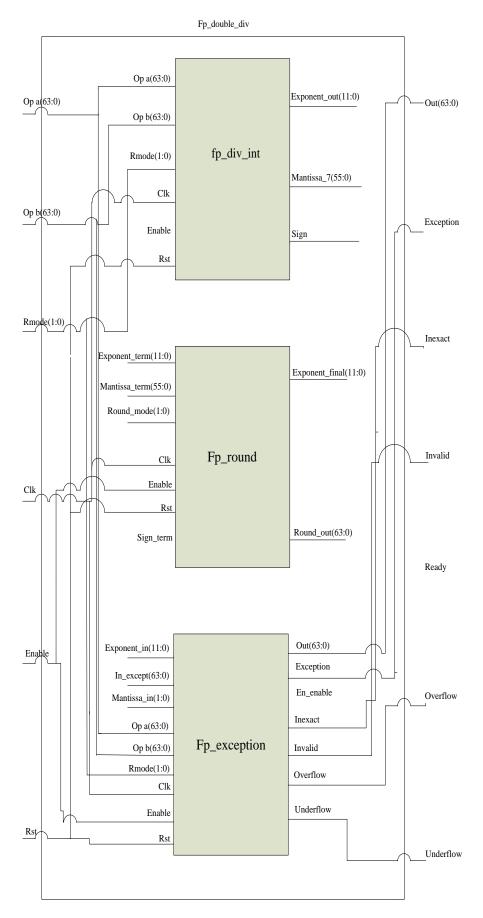


Figure 4: Inter-connection of sub modules of High Speed double precision floating point divider

The divide operation is performed in the module (fp_div_int). The leading '1' (if normalized) and mantissa of operand A is the dividend, and the leading '1' (if normalized) and mantissa of operand B is the divisor. The divide is executed long hand style, with one bit of the quotient calculated each clock cycle based on a comparison between the dividend register (dividend_reg) and the divisor register (divisor_reg). If the dividend is greater than the divisor, the quotient bit is '1', and then the divisor is subtracted from the dividend, this difference is shifted one bit to the left, and it becomes the dividend for the next clock cycle. If the dividend is less than the divisor, the dividend is shifted one bit to the left, and then this shifted value becomes the dividend for the next clock cycle.

The exponent for the divide operation is calculated from the exponent fields of operands A and B. The exponent of operand A is added to 1023, and then the exponent of operand B is subtracted from this sum. The result is the exponent value of the output of the divide operation. If the result is less than 0, the quotient will be right shifted by the amount. The divide operation takes 54 clock cycles to complete, as it takes 1 clock cycle to calculate each of the 54 bits of the quotient. The register (count_out) counts down from 53 to 0, and when it reaches 0, the 54-bit quotient register has its final value. The value that is passed on to the rounding module is stored in the 56-bit register (mantissa_7). The first most significant bit is a '0' to hold a value in case of overflow in the rounding stage, the next bit is the leading '1' for normalized numbers, and the next 52 bits are the mantissa bits. The remaining 2 bits are extra bits for rounding purposes. The first extra bit is the last bit that was calculated in the quotient. The quotient has 54 bits, while the mantissa and leading '1' are only 53 bits, so the extra bit is saved and passed on to the rounding stage. The second extra bit is calculated by performing an OR on all of the remainder bits that were leftover after the last compare between the dividend and divisor registers.

3. SIMULATION RESULTS

The simulation results of double precision floating point divider are shown in figure 5. Here the inputs opa, opb and the outputs for addition and subtraction as shown in below.

Input A: 5.0000000000e+000 Input B: 2.5000000000e+000

Enable = 1'b1;

Opa =

Opb =

Output:

The Device utilization summery of high speed double precision floating point divider is shown in figure 6. Table 1 and 2 shows the Timing Summary and Area and Frequency Comparison between the High Speed Double Precision Floating Point divider and reference paper [1]. The high speed double precision floating point divider targeting on Virtex-6 xc6vlx75t-3ff484 with a Frequency of 344.89 MHz, it required area is 653 slices.

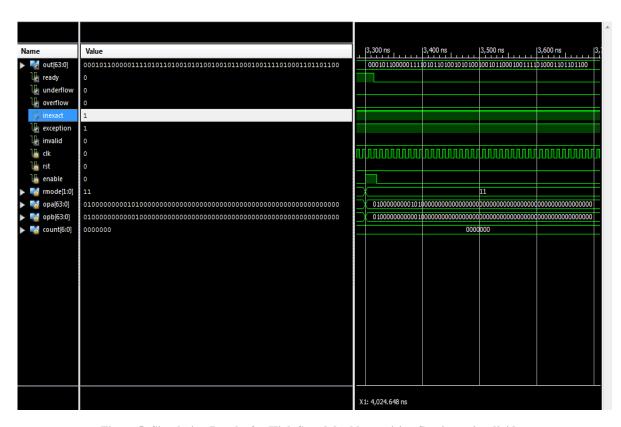


Figure 5: Simulation Results for High Speed double precision floating point divider

Device Utilization Summary			
Slice Logic Utilization	Used	Available	Utilization
Number of Slice Registers	1,980	93,120	2%
Number used as Flip Flops	1,980		
Number used as Latches	0		
Number used as Latch-thrus	0		
Number used as AND/OR logics	0		
Number of Slice LUTs	2,105	46,560	4%
Number used as logic	1,975	46,560	4%
Number using O6 output only	1,371		
Number using O5 output only	64		
Number using O5 and O6	540		
Number used as ROM	0		
Number used as Memory	1	16,720	1%
Number used as Dual Port RAM	0		
Number used as Single Port RAM	0		
Number used as Shift Register	1		
Number using O6 output only	1		
Number using O5 output only	0		
Number using O5 and O6	0		
Number used exclusively as route-thrus	129		
Number with same-slice register load	122		
Number with same-slice carry load	7		
Number with other load	0		
Number of occupied Slices	653	11,640	5%
Number of LUT Flip Flop pairs used	2,341		
Number with an unused Flip Flop	913	2,341	39%
Number with an unused LUT	236	2,341	10%
Number of fully used LUT-FF pairs	1,192	2,341	50%
Number of unique control sets	18		

Figure 6 : Device utilization summery of High Speed double precision floating point divider

Table 1: Timing summary for High Speed double precision floating point divider

Sl. No	Parameter	
1	Minimum period (ns)	2.899
2	Max Frequency (MHz)	344.89
3	Minimum input arrival time before clock (ns)	1.520
4	Maximum output required time after clock (ns)	0.562

Table 2: Area and Frequency Comparison between the High Speed Double Precision Floating Point divider and reference paper in [1]

Our Floating Point divider (Virtex-6)	In [1]
653	
1980	265 MHz
344.89	
	(Virtex-6) 653 1980

4 CONCLUSIONS

The high speed double precision floating point divider supports the IEEE 754 binary interchange format, targeted on a Xilinx Virtex-6 xc6vlx75t-3ff484 FPGA. It achieved 344.89 MFLOPs which is 30% fast compared to reference paper [1]. This design occupies 653 slices which is less area compared to reference paper [1]. In view of used flip flops this design uses 1980 flip flops. This design handles the overflow, underflow and rounding mode.

5 REFERENCES

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