

Stability of a Class of Neutral Time-Delay Systems with a Robust Control

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ABSTRACT

This paper deal with the stability problem of neutral time-delay systems. Based on the Lyapunov-Krasovskii functional theory, new theorems are proposed for a type of neutral delay systems with robust time-delay control. New delay-dependent stability conditions are developed for the system without time-delay control in first time and with time-delay control in second time. Linear matrix inequality approaches are used to solve the stability problem in these cases. Numerical examples illustrate that the proposed methods are effective and lead to less conservative results.

General Terms

Stability Analysis, Time-delay systems.

Keywords

Neutral time-delay systems, stability analysis, robust time-delay control, linear matrix inequality (LMI).

1. INTRODUCTION

Neutral systems represent an interesting class of time-delay systems where dynamical systems not only depend on present and past states but also involve derivatives with delays. The presence of time-delay in state derivatives makes system behavior more similar to real comportment. Many systems can be modeled by these systems, such as population ecology, distributed networks with loss-less transmission lines, heat exchangers, and robots in contact with rigid environments.

Recently, much attention has been given to the study of neutral delay systems' theory. Moreover, time delays are often the main sources of instability. Hence, system stability and stabilization analysis are the essential step to treat system control problem. These subjects have been studied see e.g. References [1], [2], [3] [4].

Lyapunov stability is a general theory available for any differential equation. Over the past researches on time-delay systems, many useful approaches are applied to guarantee the stability or stabilization of systems. The application of Lyapunov-Krasovskii functional theory has first started for system without neither uncertainties nor control [2] [5], some robust stability conditions based on LMI approach are given. Then, the guaranteed cost control problem for neutral time delay system with feed-back control is investigated. Some papers are interested on stability and stabilization of this type of system where a linear-quadratic cost function is considered

as a performance measure for the closed-loop system [6] [7] [8] [9] [10]. In the last years, interesting works have been concerned with uncertain neutral time-delay systems stability and stabilization analysis based on Lyapunov-Krasovskii functional theory [3] [11] [4].

In the theory of time-delay systems, there are two classes of stability conditions. The first is delay-independent conditions which guarantee that a time-delay system remains stable for all non-negative values of delay elements; see for example [5][12][13][14].

The second type is called delay-dependant conditions and involve explicitly delay elements of the system and guarantee the system stability for some restricted values of the delay elements, see for example [6] [4] [15] [16][17].

Dependant or independent-delay stability conditions, in practical system, are usually obtained by using the analysis of mathematical model. Therefore, in this paper a LMI condition is proposed equivalent to the delay-dependent stability of neutral systems with time-delay. More precisely, one displays a family of LMIs of increasing size, each of them ensuring delay-independent stability. The key result is that, reciprocally, the strong delay-dependent stability ensures that the LMIs are solvable beyond a certain rank.

The aim contributions of this paper is to develop some stability analysis conditions, witch depend on delays, based on linear matrix inequality LMI approach for a neutral delay system without delayed robust control, in first time and with delayed control in second time. The application of Lyapunov-Krasovskii functional theory is the first step to conclude these conditions.

The remainder of this paper is organized as follows. In section 2, a brief review on stability analysis of neutral time-delay systems is presented. Section 3 gives stability conditions for neutral systems with robust control without delays using linear matrix inequalities. In section 4, other conditions is proposed for neutral systems with robust delayed control. Section 5 contains simulation results to show validity of theory results. Conclusions and future work are presented in Section 6.

2. STABILITY ANALYSIS OF NEUTRAL DELAY SYSTEMS

2.1 Review

The stability analysis is an essential step for a system control and a diagnosis strategy. Some conditions are derived in literature to guarantying neutral time-delay system.

It's already cited that there are two classes of stability conditions for neutral systems with time-delay. The first is a delay independent condition. Stability criteria for neutral systems with multiple time delays are presented in [5]. Using the Lyapunov second method, Park and al. establish a new delay-independent criterion for the asymptotic stability of systems. In these criteria, the derived sufficient conditions are expressed in terms of linear matrix inequalities so that the criteria are less conservative.

In [12], sufficient conditions for the existence of these observers are derived. Using the linear matrix inequality and the linear matrix equality (LME) formulation, independent of delays stability criteria are derived in [12] for proposed observers.

Wang and al. [13] consider the H_∞ dynamic for linear neutral time-delay systems output feedback controller design problem. The approach here is based on Lyapunov functional due to Krasovsii. A sufficient condition is deduced in terms of linear matrix inequalities.

The second conditions are dependent on a delay size.

In fact, in [6], Sun and al. introduce a new form of the Lyapunov functional that contains a triple integral term $\int_{-\tau}^0 \int_{\theta}^0 \int_{t+\lambda}^t \dot{x}^T(s) R \dot{x}(s) ds d\lambda d\theta$. Two integral inequalities are used to derive a new delay-dependent stability criterion without introducing any free-weighting matrices. Using this criterion, a method of designing a stabilizing state feedback controller is also presented.

The paper of Xin and al. [17] deals with the delay-dependent stability criterion and the state observers design problem as well as observer-based stabilization problem for linear neutral delay systems. A delay-dependent stability criterion is developed, which is presented in terms of a feasibility positive definite solution to a linear matrix inequality.

2.2 Problem Formulation and preliminaries

In this paper, a stability condition for a type of neutral systems with delayed control will be presented.

Consider the following neutral time-delay system with a robust control:

$$\begin{cases} \dot{x}(t) = Ax(t) + A_h x(t-h) + A_d \dot{x}(t-d) + Bu(t) + B_1 u(t-\tau_1) \\ y(t) = Cx(t) \\ x(t) = \varphi(t), \quad t \in [-\tau, 0] \end{cases} \quad (1)$$

where $x(t) \in \mathbb{R}^n$, $u \in \mathbb{R}^m$ and $y(t) \in \mathbb{R}^p$. $h > 0, d > 0$ and $\tau_1 > 0$ are respectively the state and its derivative delay. A, A_h, A_d, B, B_1 and C are constant matrix.

$\tau = \max\{h, d\}$ and $\varphi(t) \in \mathbb{R}^n$ is the continuous function of initial condition in $[-\tau, 0]$.

Using the Newton-Lebuniz formula,

$$x(t-h) = x(t) - \int_{-h}^0 \dot{x}(t+\alpha) d\alpha$$

system (1) become as the following form:

$$\begin{cases} \dot{x}(t) = (A + A_h)x(t) - A_h \int_{-h}^0 \dot{x}(t+\alpha) d\alpha + A_d \dot{x}(t-d) + Bu(t) + B_1 u(t-\tau_1) \\ y(t) = Cx(t) \\ x(t) = \varphi(t), \quad t \in [-\tau, 0] \end{cases} \quad (2)$$

In this paper some problem will be treated:

A technique of stability analysis will be investigated for a system with a robust non delayed control including feedback control and disturbances. In second step, a stability condition is derived for a system with a robust delayed control including feedback control and disturbances.

Two known lemmas will be used in the proof of the following theorem.

Lemma 2.1 For all $\varphi(t) \in C([- \tau, 0], \mathbb{R}^n)$, the following inequality holds:

$$\|\varphi(\theta)\|^2 \leq 2\|\varphi(\theta)\|^2 + 2\tau \int_{\theta}^0 \|\dot{\varphi}(s)\|^2 ds, \quad \theta \in [-\tau, 0] \quad (3)$$

Lemma 2.2 For all $\varphi(t) \in C([- \tau, 0], \mathbb{R}^n)$ and for the system (2), there exist $\alpha_1 > 0$ and $\alpha_2 > 0$ satisfy:

$$\alpha_1 \|\varphi(0)\|^2 \leq V(\varphi(\theta), t) \leq \alpha_2 \|\varphi(\theta)\|_W^2 \quad (4)$$

3. DELAY-DEPENDENT STABILITY CRITERIA

In this section, two delay-dependent stability conditions with and without control delays are developed.

3.1 Stability of neutral delay systems without delayed control

The following theorem shows a new condition for system (2) with robust non delayed control ($B_1 = 0$).

Theorem 3.1:

Consider a neutral system (2) with a given constant $h^* > 0$ and $B_1 = 0$. The tow delays h and d are supposed different.

System (2) is asymptotically stable for any $0 < \tau \leq h^*$, if there exist matrices $X > 0$, $T > 0$, and $Y > 0$ satisfy the following LMI:

$$(5) \quad \begin{pmatrix} G(X, A, A_h) & 0 & A_d Y & B & X A^T & X C^T & X & A_h Y \\ * & -T & 0 & 0 & T A_h^T & 0 & 0 & 0 \\ * & * & -Y & 0 & Y A_d^T & 0 & 0 & 0 \\ * & * & * & I & B^T & 0 & 0 & 0 \\ * & * & * & * & -\frac{1}{1+h^*} Y & 0 & 0 & 0 \\ * & * & * & * & * & -I & 0 & 0 \\ * & * & * & * & * & * & -T & 0 \\ * & * & * & * & * & * & * & -\frac{1}{h^*} Y \end{pmatrix} < 0$$

where “*” and I denote respectively the transposed elements in the symmetric position and identity matrix.

$$G(X, A, A_h) = (A + A_h)X + X(A + A_h)^T$$

Proof:

Let us choose a Lyapunov function for the system (2) as:

$$V(x, t) = x^T(s)Px(s) + \int_{t-h}^t x^T(s)Sx(s)ds + \int_{t-d}^t \dot{x}^T(s)H\dot{x}(s)ds + \int_{-h}^0 \left(\int_{t+\alpha}^t \dot{x}^T(s)H\dot{x}(s)ds \right) d\alpha$$

where: $P > 0$, $S > 0$ and $H > 0$ are symmetric positive-definite matrices.

First, we'll start to prove that:

$$\|\varphi(\theta)\|^2 \leq 2\|\varphi(0)\|^2 + 2\tau \int_{\theta}^0 \|\dot{\varphi}(s)\|^2 ds, \quad \theta \in [-\tau, 0] \quad (5)$$

Since $\varphi(\theta) = \varphi(0) - \int_{\theta}^0 \dot{\varphi}(s)ds$, $\theta \in [-\tau, 0]$, inequality (5) can be written as:

$$\begin{aligned} \|\varphi(\theta)\|^2 &\leq \left(\|\varphi(0)\| + \int_{\theta}^0 \|\dot{\varphi}(s)\|ds \right)^2 \\ &\leq 2\|\varphi(0)\|^2 + 2 \left(\int_{\theta}^0 \|\dot{\varphi}(s)\|ds \right)^2 \\ &\leq 2\|\varphi(0)\|^2 + 2 \int_{\theta}^0 \|\dot{\varphi}(s)\|^2 ds \int_{\theta}^0 1ds \\ &= 2\|\varphi(0)\|^2 - 2\theta \int_{\theta}^0 \|\dot{\varphi}(s)\|^2 ds \\ &\leq 2\|\varphi(0)\|^2 + 2\tau \int_{\theta}^0 \|\dot{\varphi}(s)\|^2 ds \end{aligned}$$

Second, for a system (2) with Lyapunov functional candidate, there exist $\alpha_1 > 0$ and $\alpha_2 > 0$ satisfy

$$\alpha_1 \|\varphi(0)\|^2 \leq V(\varphi(\theta), t) \leq \alpha_2 \|\varphi(\theta)\|_W^2 \quad (6)$$

$$\text{where } \|\varphi(\theta)\|_W = \left(\|\varphi(0)\|^2 + \int_{-\tau}^0 \|\dot{\varphi}(\theta)\|^2 d\theta \right)^{1/2}$$

The Lyapunov functional candidate is:

$$\begin{aligned} V(\varphi(\theta), t) &= \varphi^T(\theta)P\varphi(\theta) + \int_{t-h}^t \varphi^T(\theta)S\varphi(\theta)d\theta \\ &\quad + \int_{t-d}^t \dot{\varphi}^T(\theta)H\dot{\varphi}(\theta)d\theta + \int_{-h}^0 \left(\int_{t+\alpha}^t \dot{\varphi}^T(\theta)H\dot{\varphi}(\theta)d\theta \right) d\alpha \\ V(\varphi(\theta), t) &\geq \varphi^T(\theta)P\varphi(\theta) \\ &\geq \lambda_m(P) \|\varphi(0)\|^2 \end{aligned}$$

where $\lambda_m(P)$ the smallest eigenvalues of matrix P .

Then, $\alpha_1 = \lambda_m(P) > 0$

In second hand,

$$\begin{aligned} V(\varphi(\theta), t) &\leq \lambda_M(P) \|\varphi(\theta)\|^2 + \lambda_M(S) \int_{-h}^0 \|\varphi(\theta)\|^2 d\theta + \\ &\quad \lambda_M(H) \int_{-d}^0 \|\dot{\varphi}(\theta)\|^2 d\theta + \lambda_M(H) \int_{-d}^0 \left(\int_{\alpha}^0 \|\dot{\varphi}(\theta)\|^2 d\theta \right) d\alpha \\ &\leq \lambda_M(P) \|\varphi(\theta)\|^2 + \lambda_M(S) \int_{-\tau}^0 \|\varphi(\theta)\|^2 d\theta \\ &\quad + \lambda_M(H) \int_{-\tau}^0 \|\dot{\varphi}(\theta)\|^2 d\theta + \lambda_M(H) \int_{-\tau}^0 \left(\int_{\alpha}^0 \|\dot{\varphi}(\theta)\|^2 d\theta \right) d\alpha \\ &\leq \lambda_M(P) \|\varphi(\theta)\|^2 + \lambda_M(S) \int_{-\tau}^0 \|\varphi(\theta)\|^2 d\theta \\ &\quad + (1+\tau)\lambda_M(H) \int_{-\tau}^0 \|\dot{\varphi}(\theta)\|^2 d\theta \\ &\leq 2\lambda_M(P) \|\varphi(0)\|^2 + 2\tau\lambda_M(P) \int_{\theta}^0 \|\dot{\varphi}(s)\|^2 ds \\ &\quad + \lambda_M(S) \int_{-\tau}^0 \left(2\|\varphi(0)\|^2 + 2\tau \int_{\theta}^0 \|\dot{\varphi}(s)\|^2 ds \right) \\ &\quad + (1+\tau)\lambda_M(H) \int_{-\tau}^0 \|\dot{\varphi}(\theta)\|^2 d\theta \\ &\leq 2(\lambda_M(P) + \tau\lambda_M(S)) \|\varphi(0)\|^2 \\ &\quad + (2\tau(\lambda_M(P) + \lambda_M(S)) + (1+\tau)\lambda_M(H)) \int_{-\tau}^0 \|\dot{\varphi}(\theta)\|^2 d\theta \\ &\leq \alpha_2 \left(\|\varphi(0)\|^2 + \int_{-\tau}^0 \|\dot{\varphi}(\theta)\|^2 d\theta \right) \end{aligned}$$

where

$$\alpha_2 = \max \left\{ 2(\lambda_M(P) + \tau\lambda_M(S)), \left(2\tau(\lambda_M(P) + \lambda_M(S)) + (1+\tau)\lambda_M(H) \right) \right\} > 0 \quad (7)$$

For simplicity, write $x(t) = x$, $x(t-h) = x_h$, $\dot{x}(t-d) = \dot{x}_d$.

Therefore,

Then, for all any vectors u and v , and reversible matrix θ , it is obvious that the following inequality is true.

$$\begin{aligned} (u^T \theta + \theta^{-1} v)^2 &= u^T \theta^2 u + 2u^T v + v^T \theta^{-2} v \geq 0 \\ u^T \theta^2 u + v^T \theta^{-2} v &\geq -2u^T v \end{aligned}$$

Let us consider $\theta^2 = M$, then $u^T M u + v^T M^{-1} v \geq -2u^T v$.

Hence,

$$\begin{aligned} \dot{V}(x, t) &= 2\dot{x}^T P x + x^T S x - \dot{x}_h^T S x_h + \dot{x}^T H \dot{x} - \dot{x}_d^T H \dot{x}_d \\ &\quad + \int_{-h}^0 \left[\dot{x}^T(t) H \dot{x}(t) - \dot{x}^T(t + \alpha) H \dot{x}(t + \alpha) \right] d\alpha \\ &= 2\dot{x}^T P x + x^T S x - \dot{x}_h^T S x_h + (1+h)\dot{x}^T H \dot{x} - \dot{x}_d^T H \dot{x}_d \\ &\quad - \int_{-h}^0 \dot{x}^T(t + \alpha) H \dot{x}(t + \alpha) d\alpha \\ &= x^T \left(P(A + A_h) + (A + A_h)^T P + S \right) x \\ &\quad - 2x^T P \int_{-h}^0 A_h \dot{x}(t + \alpha) d\alpha + 2x^T P A_d \dot{x}_d + 2x^T P B u \\ &\quad - \dot{x}_h^T S x_h + (1+h)\dot{x}^T H \dot{x} - \dot{x}_d^T H \dot{x}_d \\ &\quad - \int_{-h}^0 \dot{x}^T(t + \alpha) H \dot{x}(t + \alpha) d\alpha \\ &= -2x^T P \int_{-h}^0 A_h \dot{x}(t + \alpha) d\alpha - \int_{-h}^0 2x^T P A_h \dot{x}(t + \alpha) d\alpha \\ &\leq \int_{-h}^0 x^T P A_h H^{-1} A_h^T P x + \dot{x}^T(t + \alpha) H \dot{x}(t + \alpha) d\alpha \\ &\leq h x^T P A_h H^{-1} A_h^T P x + \int_{-h}^0 \dot{x}^T(t + \alpha) H \dot{x}(t + \alpha) d\alpha \end{aligned} \quad (8)$$

Substituting this inequality into (8), we obtain:

$$\begin{aligned} \dot{V}(x, t) &\leq x^T \left(P(A + A_h) + (A + A_h)^T P + S \right) x + h x^T P A_h H^{-1} A_h^T P x \\ &\quad + \int_{-h}^0 \dot{x}^T(t + \alpha) H \dot{x}(t + \alpha) d\alpha + 2x^T P A_d \dot{x}_d + 2x^T P B u - \dot{x}_h^T S x_h \\ &\quad + (1+h)\dot{x}^T H \dot{x} - \dot{x}_d^T H \dot{x}_d - \int_{-h}^0 \dot{x}^T(t + \alpha) H \dot{x}(t + \alpha) d\alpha \\ &= x^T \left(P(A + A_h) + (A + A_h)^T P + S \right) x + h x^T P A_h H^{-1} A_h^T P x \\ &\quad + 2x^T P A_d \dot{x}_d + 2x^T P B u - \dot{x}_h^T S x_h + (1+h)\dot{x}^T H \dot{x} - \dot{x}_d^T H \dot{x}_d \\ &= x^T \left(P(A + A_h) + (A + A_h)^T P + S + h P A_h H^{-1} A_h^T P \right) x \\ &\quad + 2x^T P A_d \dot{x}_d + 2x^T P B u - \dot{x}_h^T S x_h + (1+h)\dot{x}^T H \dot{x} - \dot{x}_d^T H \dot{x}_d \end{aligned}$$

We have also,

$$\begin{aligned} \dot{x}^T (H + hH) \dot{x} &= x^T A^T (1+h) H A x + 2x^T A^T (1+h) H A_h x_h \\ &\quad + 2x^T A^T (1+h) H A_d \dot{x}_d + 2x^T A^T (1+h) H B u \\ &\quad + 2x_h^T A_h^T (1+h) H A_d \dot{x}_d + 2x_h^T A_h^T (1+h) H B u \\ &\quad + x_h^T A_h^T (1+h) H A_h x_h + 2\dot{x}_d^T A_d^T (1+h) H B u \\ &\quad + \dot{x}_d^T A_d^T (1+h) H A_d \dot{x}_d + u^T B^T (1+h) H B u \end{aligned}$$

So, Lyapunov functional is obtained as:

$$\begin{aligned} \dot{V}(x, t) &\leq x^T \left(P(A + A_h) + (A + A_h)^T P + S + h P A_h H^{-1} A_h^T P \right) x \\ &\quad + 2x^T P A_d \dot{x}_d + 2x^T P B u - \dot{x}_h^T S x_h + x^T A^T (1+h) H A x \\ &\quad + 2x^T A^T (1+h) H A_h x_h + 2x^T A^T (1+h) H A_d \dot{x}_d \\ &\quad + 2x^T A^T (1+h) H B u + 2x_h^T A_h^T (1+h) H A_d \dot{x}_d \\ &\quad + 2x_h^T A_h^T (1+h) H B u + x_h^T A_h^T (1+h) H A_h x_h \\ &\quad + 2\dot{x}_d^T A_d^T (1+h) H B u + \dot{x}_d^T A_d^T (1+h) H A_d \dot{x}_d \\ &\quad + u^T B^T (1+h) H B u - \dot{x}_d^T H \dot{x}_d \end{aligned}$$

Since $y^T y - u^T u < 0$,

so,

$$\dot{V}(x, t) + y^T y - u^T u \leq \tilde{x}^T \Gamma(h) \tilde{x} < 0 \quad (9)$$

where $\tilde{x} = \begin{bmatrix} x^T & x_h^T & \dot{x}_d^T & u^T \end{bmatrix}^T$ and

$$\Gamma(h) = \begin{pmatrix} \Phi(h) & (1+h)A^T H A_h & P A_d + (1+h)A^T H A_d & P B + (1+h)A^T H B \\ * & (1+h)A_h^T H A_h - S & (1+h)A_h^T H A_d & (1+h)A_h^T H B \\ * & * & (1+h)A_d^T H A_d - H & (1+h)A_d^T H B \\ * & * & * & (1+h)B^T H B - I \end{pmatrix}$$

$$\begin{aligned} \Phi(h) &= P(A + A_h) + (A + A_h)^T P + S + h P A_h H^{-1} A_h^T P \\ &\quad + (1+h)A^T H A + C^T C \end{aligned}$$

Obviously, inequality (9) implies $\dot{V} < 0$, if $\Gamma(h) < 0$. Note that the matrix $\Gamma(h) < 0$ is monotonic increasing with respect to $\tau > 0$ in the sense of positive definiteness, therefore, $\Gamma(h) < 0$ holds for $0 < \tau \leq h^*$ if $\Gamma(h^*) < 0$ holds.

By Schur complement, $\Gamma(h^*) < 0$ holds if and only if:

$$\begin{pmatrix} G(P, S) & 0 & P A_d & P B & A^T & P A_h \\ * & -S & 0 & 0 & A_h^T & 0 \\ * & * & -H & 0 & A_d^T & 0 \\ * & * & * & -I & B^T & 0 \\ * & * & * & * & -\frac{1}{1+h^*} H^{-1} & 0 \\ * & * & * & * & * & -\frac{1}{h^*} H \end{pmatrix} < 0 \quad (10)$$

where $G(P, S) = P(A + A_h) + (A + A_h)^T P + S + C^T C$.

Multiplying (10) on both sides by $\text{diag}\{X, T, Y, I, Z, Y\}$ (I is the identity matrix) and then applying Schur complement to the result, we have

$$\begin{pmatrix}
 G(X, A, A_h) & 0 & A_d Y & B & X A^T & X C^T & X & A_h Y \\
 * & -T & 0 & 0 & T A_h^T & 0 & 0 & 0 \\
 * & * & -Y & 0 & Y A_d^T & 0 & 0 & 0 \\
 * & * & * & I & B^T & 0 & 0 & 0 \\
 * & * & * & * & -\frac{1}{1+h^*} Y & 0 & 0 & 0 \\
 * & * & * & * & * & -I & 0 & 0 \\
 * & * & * & * & * & * & -T & 0 \\
 * & * & * & * & * & * & * & -\frac{1}{h^*} Y
 \end{pmatrix} < 0
 \quad (11)$$

where $G(X, A, A_h) = (A + A_h)X + X(A + A_h)^T$

3.2 Stability of neutral delay systems with delayed control

Some neutral delay system models contain time delay in control. In this section, a stability condition is derived for system (2) with delay robust control ($B_1 \neq 0$). In theorem 3.2, stability criterion is created in term of linear matrix inequality.

Theorem 3.2:

Consider a neutral system (2) with given constants $h^* > 0$, h , d and τ_1 are supposed different. The system (2) is asymptotically stable for any $0 < \tau \leq h^*$, if there exist matrices $X > 0$, $T > 0$, and $Y > 0$ satisfy the following LM

$$\begin{pmatrix}
 G(X, A, A_h) & 0 & A_d Y & B & B_1 & X A^T & X C^T & X & A_h Y \\
 * & -T & 0 & 0 & 0 & T A_h^T & 0 & 0 & 0 \\
 * & * & -Y & 0 & 0 & Y A_d^T & 0 & 0 & 0 \\
 * & * & * & I & 0 & B^T & 0 & 0 & 0 \\
 * & * & * & * & I & B_1^T & 0 & 0 & 0 \\
 * & * & * & * & * & -\frac{1}{1+h^*} Y & 0 & 0 & 0 \\
 * & * & * & * & * & * & -I & 0 & 0 \\
 * & * & * & * & * & * & * & -T & 0 \\
 * & * & * & * & * & * & * & * & -\frac{1}{h^*} Y
 \end{pmatrix} < 0
 \quad (12)$$

where “*” and I denote respectively the transposed elements in the symmetric position and the identity matrix.

$$G(X, A, A_h) = (A + A_h)X + X(A + A_h)^T$$

Proof:

Consider the vector $U(t)$ and the matrix \tilde{B} expressed as:

$$U(t) = \begin{pmatrix} u(t) \\ u(t - \tau_1) \end{pmatrix} \text{ and } \tilde{B} = \begin{pmatrix} B & B_1 \end{pmatrix}.$$

The system (2) becomes

$$\begin{cases}
 \dot{x}(t) = (A + A_h)x(t) - A_h \int_{-h}^0 \dot{x}(t + \alpha) d\alpha + A_d \dot{x}(t - d) + \tilde{B}U(t) \\
 y(t) = Cx(t) \\
 x(t) = \varphi(t), \quad t \in [-\tau, 0]
 \end{cases}
 \quad (13)$$

Hence, the result follows immediately by applying Theorem 3.1 to the system (13).

4. NUMERICAL EXAMPLE

In this section, numerical examples are presented to illustrate the effectiveness of precedent results. A first leads into system with non delay control and a second introduce the delay in the system control.

Example 1:

Consider system (1) with:

$$A = \begin{pmatrix} -1 & 1 \\ 0 & -4 \end{pmatrix}, \quad A_h = \begin{pmatrix} 0.15 & 0.05 \\ 0 & 0.1 \end{pmatrix}, \quad A_d = \begin{pmatrix} -0.05 & 0.02 \\ 0.01 & 0 \end{pmatrix}, \\
 B = \begin{pmatrix} 0 \\ 1 \end{pmatrix}, \quad C = \begin{pmatrix} 1 & 1 \end{pmatrix}, \quad h = 1 \text{ and } d = 0.9.$$

From Theorem 3.1 the following results can be obtained as:

$$X = \begin{pmatrix} 0.4688 & -0.0092 \\ -0.0092 & 0.2675 \end{pmatrix}, \quad T = \begin{pmatrix} 1.7096 & -0.0071 \\ -0.0071 & 1.6667 \end{pmatrix} \text{ and} \\
 Y = \begin{pmatrix} 1.9208 & -0.0318 \\ -0.0318 & 2.1572 \end{pmatrix}.$$

X , T , and Y are positives. Thus, the system is asymptotically stable. The following figure (Fig.1) illustrates the unit step response of system and proves the system stability.

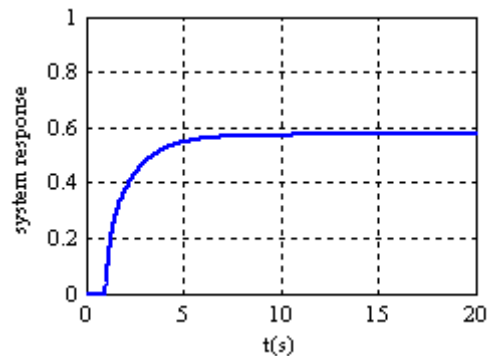


Fig.1 step response

Example 2:

Consider a system (1) with:

$$A = -1, \quad A_h = 1, \quad A_d = 1, \quad B = \begin{bmatrix} 1 & -1 \end{bmatrix}, \quad B_1 = \begin{bmatrix} -1 & 1 \end{bmatrix}, \quad C = 1.$$

$h = 1$, $d = 0.9$ and $\tau_1 = 0.5$.

By Theorem 3.2, the feasibility of the LMI (12) is obtained with:

$$X = 0.3156, \quad Y = 0.2180, \quad Z = 0.2796.$$

X , T , and Y are positives. Hence the system is stable. The figure (Fig.2) illustrate a system response when the control is

a unit step delayed with $\tau_1 = 0.5$. The figure shows that the system is asymptotically stable.

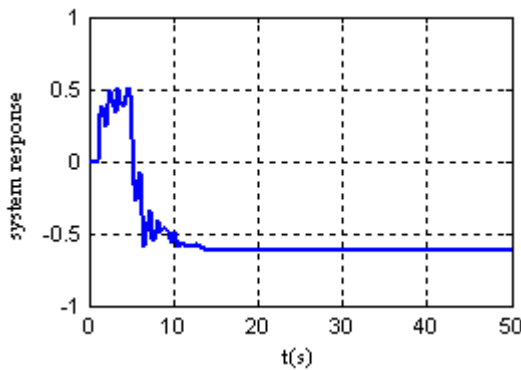


Fig.2 step response of system

5. CONCLUSION

In this paper, the stability analysis problem for a class of neutral delay system with robust control has been considered. The Lyapunov stability theory and LMI have been used to guarantee the delay-dependent robust stability for the system. Moreover, the problem of delay in the robust control for neutral system has been investigated. Finally, a numerical example is illustrated to show the validity of the proposed methods. Based on these results, the problem of control or diagnosis can be achieved in the future for real example of neutral time-delay.

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