

H_∞ norm Computation for Fractional Order Control System

Suman Suri
RamraoAdik Institute of
Technology,
Nerul, Navi Mumbai

Mukesh D. Patil
RamraoAdik Institute of
Technology,
Nerul, Navi Mumbai

Vishwesh A. Vyawahare
Systems and Control Engg.
Indian Institute of Technology,
Bombay, Mumbai

ABSTRACT

In this paper, we propose two tools for fractional systems H_∞ norm computation; one is based on an extension of widely used tools designed for integer systems that is by manipulating properties of singular values. The other is the computing H_∞ norm graphically as peak value of bode magnitude plot.

General Terms

H_∞ norm, Fractional systems

Keywords

Fractional order system (FOS),

1. INTRODUCTION

In recent years there has been an increasing attention to fractional-order systems. These systems are of interest for both modelling and control purposes. In general, Fractional differentiation is now a well known tool for controller synthesis. Several presentations and applications of the fractional PID controller [1], [2], [3] [4] and of CRONE control [5] demonstrate their efficiency. Fractional differentiation also permits a simple representation of some high order complex integer systems [6]. Consequently, basic properties of fractional systems have been investigated these last ten years and criteria and theorems are now available in the literature concerning stability [7], observability, and controllability [8] of fractional systems. Thus, several studies have been made on FOS stability and the most well known stability criteria is Matignon's criteria which enables to test systems stability through the location of the state matrix Eigen values in the complex plane.

The H_∞ norm reflects how much a dynamic system amplifies or attenuates its input at the frequency at which the amplification is maximal. This control technique may be applied to both SISO (single-input, single-output) and MIMO (multiple-input, multiple-output) plants, and its results achieve a remarkable robustness (Lublin *et al.*, 1996; Doyle *et al.*, 1989)[11]. By computing H_∞ norm we continue to explore the use of frequency domain techniques for design of feedback

"Permission to make digital or hard copies of part or all of this work for personal or classroom use is granted without fee provided that copies are not made or distributed for profit or commercial advantage and that copies bear this notice and the full citation on the first page. Copyrights for components of this work owned by others than IJCA must be honored. Abstracting with credit is permitted. To copy otherwise, to republish, to post on servers or to redistribute to lists, needs an acknowledgement to IJCA."

systems. H_∞ control is strongly linked to the weighted sensitivity functions. Performance specification is then of great importance in H_∞ control approach as means of loop shaping [9].

In this paper, we propose two tools for fractional systems H_∞ norm computation; one is based on an extension of widely used tools designed for integer systems that is by manipulating properties of singular values [13]. The other is the computing H_∞ norm graphically as peak value of bode magnitude plot.

2. NOTATIONS AND DEFINITIONS

2.1 Fractional Calculus

Riemann-Liouville [10] fractional differentiation is used and the fractional integral of a function $f(t)$ is defined by

$$D^{-\alpha}f(t) = D^m J^{m-\alpha}f(t) = \frac{d^m}{dt^m} \left[\frac{1}{\Gamma(m-\alpha)} \int_0^t (t-\tau)^{\alpha-m-1} f(\tau) d\tau \right]$$

where

$$\Gamma(x) = \int_0^\infty t^{x-1} \exp^{-t} dt$$

is the Gamma function which is an important special function in fractional calculus. It is an extension of the factorial function, that is, if n is a positive integer.

$$\Gamma(n) = (n-1)!$$

2.2 LTI COMMENSURATE FRACTIONAL ORDER SYSTEMS

In case of commensurate-order system [9], the transfer function is given by

$$G(s) = \frac{\sum_{k=0}^m b_k (s^\alpha)^k}{\sum_{k=0}^n a_k (s^\alpha)^k}$$

which can be considered as pseudofractional function, $H(\lambda)$, of the variable $\lambda = s^\alpha$, as in equation

$$\lambda = s^\alpha = \frac{\sum_{k=0}^m b_k \lambda^k}{\sum_{k=0}^n a_k \lambda^k}$$

The pseudo state space representation is of the form

$$D^\alpha = Ax(t) + Bu(t) \quad (1)$$

$$y(t) = Cx(t) + Du(t)$$

where $x(t) \in R^n$ is the pseudo state vector, $u(t) \in R^m$ is the input vector, $y(t) \in R^p$ is the output vector, α is the fractional order of the system and A, B, C and D are constant matrices. D^α is the fractional differentiation operator of order α and the transfer

function representation for fractional order is given as in equation

$$G(s) = C(s^\alpha I - A)^{-1}B + D$$

2.3 Stability

Stability analysis for fractional-order systems is difficult by simply examining the characteristic equation, either by finding its dominant roots or by using algebraic methods. LTI IOS stability can be checked via the location of the state matrix eigenvalues in the complex plane. This result was extended to LTI commensurate FOS of order $0 < \nu < 1$ by D. Matignon.

Theorem 1 (Matignon, 1996 [7]): System, with minimal triplet (A, B, C) and $0 < \nu < 1$, is BIBO stable if

$$|\arg(\text{eig}(A))| > \alpha \frac{\pi}{2}$$

This result remains valid when $1 < \nu < 2$. Stability domain is thus defined as follow

$$D_s = \left\{ z \in \mathbb{C} \mid \arg(\text{eig}(A)) \right\} > \alpha \frac{\pi}{2} \left. \right\}$$

3. H_∞ NORM FOR FRACTIONAL ORDER SYSTEMS (FOS)

As for LTI IOS, the H_∞ norm of stable FOS from its transfer function $G(s)$ as follows:

Definition 1: H_∞ norm of stable FOS system (1) is:

$$\|G(s)\|_\infty \triangleq \max_{\omega \in \mathbb{R}} \bar{\sigma}(G(j\omega)),$$

where $\bar{\sigma}(G(j\omega))$ is the largest singular value $(G(j\omega))$ at frequency ω :

$$\begin{aligned} \bar{\sigma}(G(j\omega)) &= \max_{i=\{1 \dots \min\{m,p\}\}} \sigma_i(G(j\omega)) \\ &= \max_{i=\{1 \dots \min\{m,p\}\}} \sqrt{\lambda_i(G(j\omega))^*(G(j\omega))} \end{aligned} \quad (2)$$

Steady state response of FOS (1) to sinusoidal input $u(j\omega)$ is $y(j\omega) = G(j\omega)u(j\omega)$. At frequency ω , the gain $\frac{\|y(\omega)\|_2}{\|u(\omega)\|_2}$, depending on vector $u(j\omega)$ is:

$$\bar{\sigma}(G(j\omega)) = \max_{u(j\omega) \neq 0} \frac{\|y(j\omega)\|_2}{\|u(j\omega)\|_2}$$

Worst case frequency gain is thus given by H_∞ norm of FOS:

$$\|G(s)\|_\infty = \max_{\omega \in \mathbb{R}} \max_{u(j\omega) \neq 0} \frac{\|y(j\omega)\|_2}{\|u(j\omega)\|_2}$$

In time domain, above equation can be written as:

$$\|G(s)\|_\infty = \max_{u(t) \neq 0} \frac{\|y(t)\|_2}{\|u(t)\|_2} = \max_{\|u(t)\|_2=1} \|y(t)\|_2$$

Therefore, H_∞ norm can be interpreted in time domain as the largest energy among output signals resulting from all inputs of unit energy. Consequently, H_∞ norm physical interpretation, in frequency and time domains, is the same for FOS as for IOS.

4. COMPUTATION OF FRACTIONAL SYSTEM H_∞ NORM BASED ON SINGULAR VALUES

Definition 1 and relation (2) imply that H_∞ norm of FOS is less than γ if:

$$\forall \omega \in \mathbb{R}, \max_{i=\{1 \dots \min\{m,p\}\}} \sqrt{\lambda_i(G(j\omega))^*(G(j\omega))} < \gamma \quad (3)$$

where γ denotes a real positive number satisfying

$$\gamma > \sigma_{\max}(D).$$

H_∞ norm of the FOS described in (1) is bounded by γ and equation (3) can be written as

$$\forall \omega \in \mathbb{R}, \lambda_i \left((G(j\omega))^*(G(j\omega)) \right) < \gamma^2$$

Due to Eigenvalue properties, above relation can be rewritten as:

$$\forall \omega \in \mathbb{R}, \lambda_i \left((\gamma^2 I - G(j\omega))^*(G(j\omega)) \right) > 0$$

which is equivalent to the Linear Matrix Inequality (LMI):

$$\forall \omega \in \mathbb{R}, \lambda_i \left((\gamma^2 I - (G(j\omega))^*(G(j\omega))) \right) > 0$$

Above equation is satisfied if and only if

$\forall \omega \in \mathbb{R}, (\gamma^2 I - (G(j\omega))^*(G(j\omega)))$ is non singular, that is if and only if $(\gamma^2 I - (G(-s))^T(G(s)))$ has no zero on imaginary axis.

Hence, the H_∞ norm is bounded by γ if and only if system whose transfer matrix

is $G_\gamma(s) = (\gamma^2 I - (G(-s))^T(G(s)))^{-1}$ is asymptotically stable.

5. COMPUTATION OF FRACTIONAL SYSTEM H_∞ NORM BASED ON PEAK MAGNITUDE

A control engineering interpretation of the infinity norm of a scalar transfer function is the distance in the complex plane from the origin to the farthest point on the Nyquist plot. As in equation, H_∞ norm can also be evaluate as the peak value on the Bode magnitude plot a transfer function

$$\|G(s)\|_\infty = \sup_{\omega \in \mathbb{R}} |G(j\omega)|$$

5.1 Procedure

H_∞ norm of fractional system can be computed graphically based on following steps,

Step1: Check that transfer function given is *proper* (order of denominator is greater than numerator).

Step2: Prove that transfer function given is stable.

Step3: If it satisfies step1 and step2, i.e., it belongs to RH_∞ .

Step4: Locate the peak magnitude from Bode magnitude plot.

5.2 Examples

In this section, H_∞ norm of two fractional systems are calculated by using the above mentioned procedure.

5.2.1 Example 1

Podlubny Transfer Function:

$$F_1(s) = \frac{1}{0.8s^{2.2} + 0.5s^{0.9} + 1}$$

Step1: $F_1(s)$ is proper.

Step2 : $F_1(s)$ is stable as poles in higher reimann sheet are always stable, thus poles in principal reimann sheet is to verify whether they fulfil the following condition for stability,

$$\frac{\pi}{m} < |\arg(\omega)| < \frac{\pi}{2m}$$

Step2.1 Mapping, $\omega = s^{1/10}$

$$F_1(s) = \frac{1}{0.8\omega^{22} + 0.5\omega^9 + 1}$$

Step2.2 solve

$$0.8\omega^{22} + 0.5\omega^9 + 1 = 0$$

The poles that are on the principal Riemann sheet are $\omega = 1.0045 \pm 0.1684j$ as shown in Fig (1)

Step2.3: Take absolute value of the poles that are on the principal Riemann sheet which is 1.661.

Step2.4: Check the condition for stability

$$\frac{\pi}{10} < 1.661 < \frac{\pi}{20}$$

Hence, $F_1(s)$ is stable.

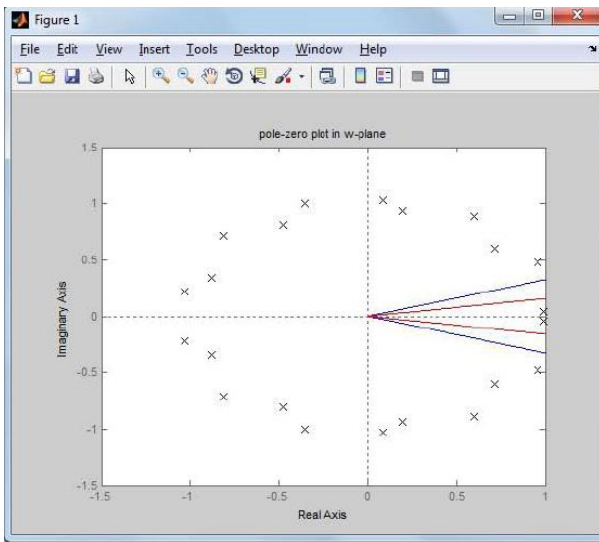


Fig 1: ω plane for $F_1(s)$

Step3: Hence, $F_1(s) \in RH_\infty$.

Step4: The peak magnitude is 13.2875db as shown in figure (2) is the H_∞ for $F_1(s)$.

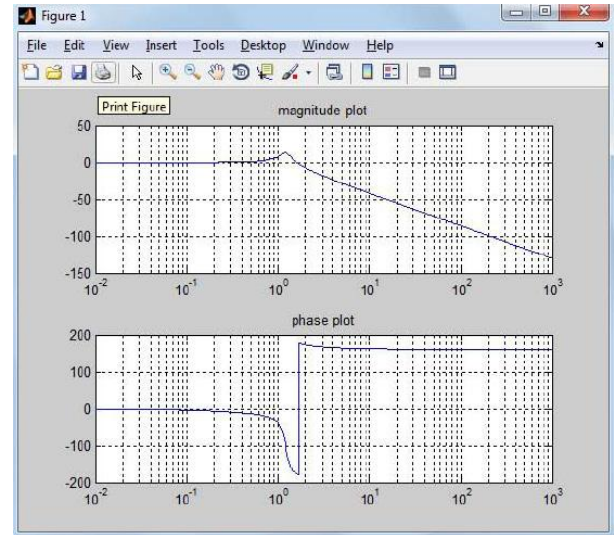


Fig 2: Bode plot for $F_1(s)$

5.2.2 Example 2

$$F_2(s) = \frac{1}{s^{2.3} + 3.2s^{1.4} + 2.4s^{0.9} + 1}$$

Step1: $F_2(s)$ is proper.

Step2 : $F_2(s)$ is stable as poles in higher reimann sheet are always stable, thus poles in principal reimann sheet is to verify whether they fulfil the following condition for stability,

$$\frac{\pi}{m} < |\arg(\omega)| < \frac{\pi}{2m}$$

Step2.1 Mapping, $\omega = s^{1/10}$

$$F_2(s) = \frac{1}{\omega^{23} + 3.2\omega^{14} + 2.4\omega^9 + 1}$$

Step2.2 solve

$$\omega^{23} + 3.2\omega^{14} + 2.4\omega^9 + 1 = 0$$

The poles that are on the principal Riemann sheet are $\omega = 0.0149 \pm 0.1906j$ as shown in Fig (3)

Step2.3: Take absolute value of the poles that are on the principal Riemann sheet which is 0.1912.

Step2.4: Check the condition for stability

$$\frac{\pi}{10} < 0.1912 < \frac{\pi}{20}$$

Hence, $F_2(s)$ is stable.

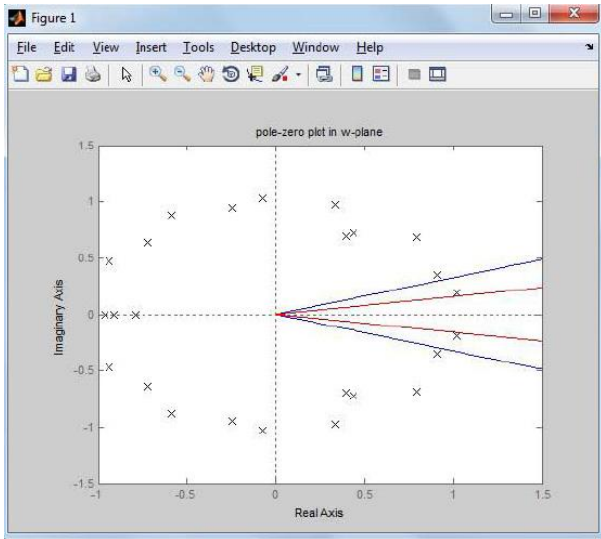


Fig 3: w plane for $F_2(s)$

Step3: Hence, $F_2(s) \in RH_{\infty}$.

Step4: The peak magnitude is 94.4323dbas shown in figure (4) is the H_{∞} for $F_2(s)$.

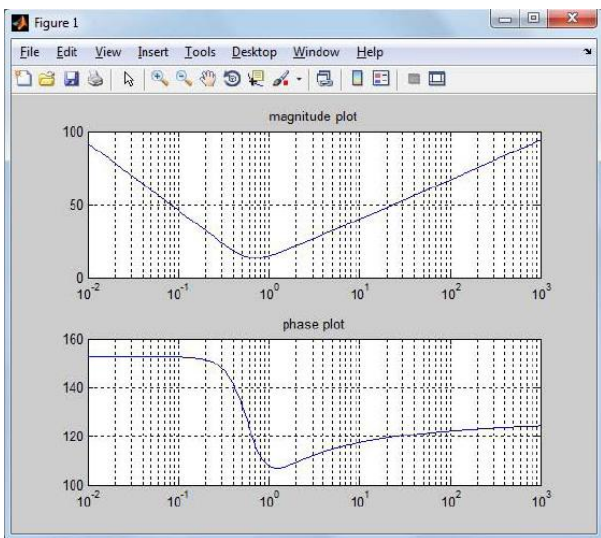


Fig 4: Bode plot for $F_2(s)$

6. ACKNOWLEDGMENTS

I am highly indebted to both the honourable guides Prof. M.D.Patil and Prof. VishweshVyawahare for their guidance and constant supervision as well for their immense support.

7. CONCLUSION

Fractional PID regulator and CRONE [8] robust regulators are now well known in the field of fractional differentiation application in control theory. H_{∞} norm plays a vital role in designing an efficient controller and to handle systems with uncertainties and disturbances and with high performance. In order to develop control methods for more complex fractional

systems than the linear one, this paper proposes two tools for the computation of a fractional system H_{∞} norm. The first one is based on singular value properties of an integer order system and the other is based on peak magnitude plot of bode plot. Computing the H_{∞} norm with these methods includes stability analysis of a fractional system and can be easily implemented. Our goal is now to design a H_{∞} controller for fractional order system.

8. REFERENCES

- [1] Monje, C., A., Vinagre, B., M., Chen, Y., Q., Feliu, V., Lanusse, P., Sabatier, J., "Proposals for fractional PID tuning", *FDA 04*, Bordeaux, France, 2004.
- [2] Podlubny, I., 1999, "Fractional-Order systems and PID-Controllers", in *IEEE Trans. on Aut. Cont.*, vol. 44, no. 1, pp. 208-214.
- [3] Caponetto, R., Fortuna, L., Porto, D., "A new tuning strategy for a non integer order PID controller", *FDA 04*, Bordeaux, France, 2004
- [4] Chen, Y., Q., Moore, K., L., Vinagre, B., M., Podlubny, I., "Robust PID controller auto tuning with a phase shaper", *First IFAC workshop on Fractional Derivative and its Application*, *FDA 04*, Bordeaux, France, 2004.
- [5] Matignon, D., July 1996, "Stability results on fractional differential equations with applications to control processing", in *Computational Engineering in Systems and Application multiconference*, vol. 2, pp.963-968, IMACS, IEEE-SMC.
- [6] Doyle and Stein. Robustness with observer. volume AC 24-4, pages 607–611. IEEE Transactions on Automatic Control, 1979.
- [7] Battaglia, J-L., Cois, O., Puissegur, L., Oustaloup, A., "Solving an inverse heat conduction problem using a non-integer identified model", *International Journal of Heat and Mass Transfer*, Vol 44, pp 2671-2680, juillet 2001.
- [8] Mathieu, B Oustaloup, A., Mathieu, B., 1999, *La commande CRONE du scalaire au multivariable*, Hermes Science Publications, Paris.
- [9] Jocelyn Sabatier Lamine Fadiga, Christophe Farges and Mathieu Moze. "On computation of H_{∞} norm for commensurate fractional order systems". pages 12–15, Orlando, FL, USA, December 2011. IEEE Conference on Decision and Control and European Control Conference (CDC-ECC).
- [10] K.B.Oldham and J.Spanier. *The Fractional Calculus*. Academic press, New York.
- [11] John C. Doyle, Keith Glover, and Khargonekar. "State-space solutions to standard H_2 and H_{∞} control problems". volume 34(8). IEEE Transactions on Automatic Control,
- [12] Ivo Petras. *Fractional Order Nonlinear System*. Springer, 2011.
- [13] Gahinet P. and Apkarian P. The linear matrix inequality approach to h_{∞} control, volume 4, pages 421–448. Int. J. Robust and Nonlinear Control, 1994.