Different Parameter Analysis of CMOS Charge Sharing Latch Comparator using 90nm Technology

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ABSTRACT

In the present works, mainly power consumption of circuit is main issue for every designers. This paper mainly dealing with the implementation of CMOS Charge sharing latch Comparator and analysis of it using different parameter in 90nm. For the implementation of new design the features of two important Comparator are combined so that the power dissipation is reduced and speed of new design is increased. These two Comparator are Resistive Dividing Comparator and Differential Current sensing respectively. The simulation result is shown in 90nm technologies, for 2.4GHz clocked comparator by using 0.9V input. Again the result of 90nm technology is compared with 180nm technology to show the reduction of power and enhancement of speed. The implementation is done on the Cadence virtuoso design environment.

Keyword: Power analysis, CMOS latch comparators.

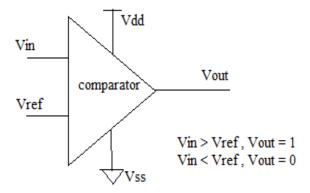
1. INTRODUCTION

Now a days, its requires the use of very low devices in the circuits and systems. Low power A/D converters ,sense amplifier and data converter, which is used for many application is using comparator as its basic element which causes reduction in power consumption by removing preamplifying stage. Due to removal of many pre amplifier stages area also get reduced. Voltage Comparator are generally used as a first stage of ADC design and output of the Comparator is used as time reference to mark the input value threshold crossing of the circuit.

In comparator design it uses back to back inverter and positive feedback mechanism for the convertion of small input voltage to full scale digital value. For such type of comparator offset voltages due to device mismatch like threshold voltages, current factor, internal/external capacitance is main limitation.

To eliminate this limitation comparator uses pre-amplifier stages before latch so that correct output can be produced. But pre-amplifier stages also has some disadvantages that it consume more static power and reduced gain due to varying technology scale, and latch comparator without pre-amplifier stage is desirable with using digital calibration technique.

Comparator is a circuit which compares the Analog signal with reference signal and produces Digital full scale output according to the comparisons



Above diagram shows the basic operation of Comparator circuit

In this paper it is discussed charge sharing latch comparator which shows power improvement compare to comparator using pre-amplifier stages. Here two main comparators are discussed below which good features are combined and used for new charge sharing topology. Let starts our discussion form two basic circuits, before discussing new design [1][2][3][4].

1.1 Resistive Dividing Comparator

This is the basic circuit of Comparator which is used for the new design implementation. This circuit is simple and more easily can be analysed. Here the main problem is more power consumption for the resistive dividing circuit in reset mode which causes power dissipation.[5],[6].

The circuit operation is given below and fig.1 shown for the resistive dividing Comparator.

Circuit works in two phases:

1. Reset phase 2.Regeneration phase.

Reset phase:

When the clk is low circuit comes to reset phase. Here the M11 get turn off, and circuit is cut off from the ground. Precharging of circuit starts working as transistor M1 and M2 get turn-on which causes the output to charge to Vdd. And Vdd charges both the output capacitor which result in power consumption.

Regeneration phase:

When clk is high it started to come in regeneration phase, and M11 get turn on.M1 and M2 turns off. So now the circuit starts comparing the two input voltages.

Advantages of this comparator are it is easy to adjust the threshold voltage by varying the transistor width instead of changing any of reference voltages.

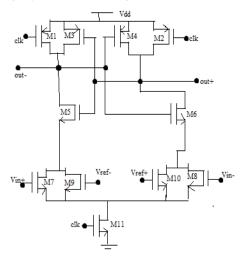


Fig.1: Resistive Dividing Comparator

1.2 Differential Current Sensing comparator:

The main advantage of this comparator is that its comparing circuit convert the input voltages to differential current at the output. Mainly it uses M12 transistor which connect both output together to share the charges instead of pre-charging to Vdd. Given below is the circuit diagram in fig.2.

During Reset phase output charge will drained off through transistor M4-M5. Now during regeneration phase when clk is low then M12 get turn off and M1 is on. This starts latch circuit to work in regeneration mode. When the output is higher than threshold voltage M6 and M7 gets turn on and its connected the output to the input comparing circuit. At the end output with low conductance change to Vdd and high conductance is change to ground.[6],[7].

A disadvantage of circuit is that it consumes more power to drive current at regeneration phase by both output.

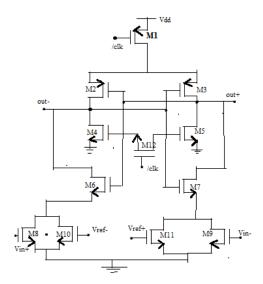


Fig.2: Differential Current Sensing Comparator

1.3 Implemented new design of Comparator

This is the new implemented design circuit which is used for the consideration and its simulation result is performed and analysed and compared in both technology 180nm and 90nm. The good properties of above two comparators are used in this design which is pass transistor which share the charges and width is variable to reduce threshold voltages. This greatly reduces the power consumption both at reset and regeneration phase by cut-off of pre-charging circuit and sharing charges at both load capacitors which makes output not to go below threshold voltage so that input can be compare faster during regeneration interval.

The operation of the circuit is as follows.

When the clock is low the circuit is in the reset mode, during this stage latch circuits of comparator get disconnected from Vdd and Ground due to disconnection of transistor M11 and M10. Hence in comparator circuit no pre-charging Pmos stage and Nmos transistor M7 as pass-transistor is used to share the charge of output voltages which is equalised to Vdd/2. This improves the power of the circuit which is half reduction compares to the circuit having pre-charging circuit. It improves the speed and it get faster. Speed can be again increased by changing the width of transistor M4 and M6, mainly increasing the width reduces charge lost. This improves the power consumption of the circuit. When it apply input signal Vin which is higher compare to reference voltages, then high Output is resulted from these circuit likewise if input signal has value lower compare to the reference voltages, Output becomes low.

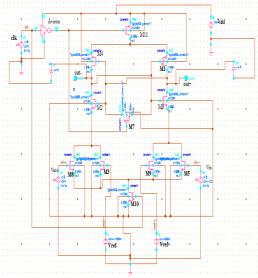


Fig.3: Charge sharing latch comparator

2. SIMULATION RESULTS

Design has been simulated in both 180nm and 90nm process technologies and the results are compared with those in old paper[6]. Lower power consumption and high operational speed is most useful part of this design.

2.1 180nm technology

Table.1 CMOS technology parameter

	All Pmos	All Nmos
Width	6000nm	3000nm
Length	180nm	180nm

2.1.1 Dc analysis: Gain: 25.61dB

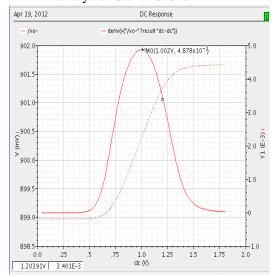


Fig.4: Dc waveform between input and output voltages

2.1.2 Transient analysis

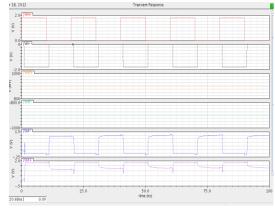


Fig.5: Transient waveform

2.1.3 Analysis of Noise Signal:

Output Noise value: $9.97\mu V/sq(Hz)$

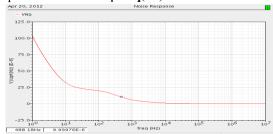


Fig.6: Output Noise Waveform

2.2 90nm technology:

Table.2 CMOS technology parameter

	All PMOS	All NMOS
Width	240 nm	120nm
Lenght	100nm	100nm

2.2.1 Output voltage

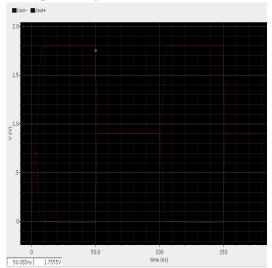


Fig.7: Charge sharing output voltage waveform

2.2.2 Transient analysis

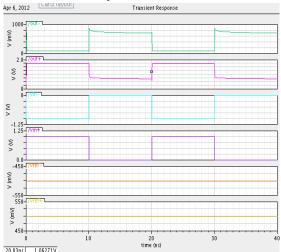


Fig.8: Transient waveform of charge sharing comparator

3. ANALYSIS USING DIFFERENT PARAMETERS

3.1 Technology Effect:

In these paper 90nm technology is used which having channel length smaller than the 180nm technology. This improves speed of the design.

3.2 Transistor width Effect:

For this paper it is used 6000nm width Pmos and 3000nm width Nmos in 180nm technology and 240nm for Pmos and 120nm width for Nmos in 90nm technology. So the new design showed speed decrement compared to 180nm, but the area of the design is smaller. From the below waveforms it is shown that as width increases the delay of the circuit decreases. As width increases from 6 μ m to 10 μ m delay line increases from line 1 to 5 shown below in fig.9.Same result also is for fig.10.

180nm technology

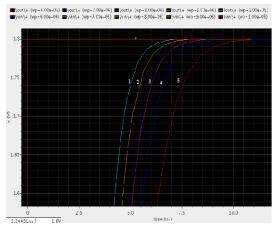


Fig.9: Simulation result for Different value of width in 180nm

90nm technology

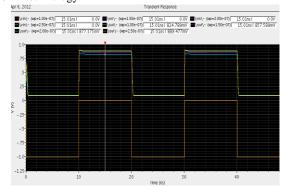


Fig.10: Simulation result for Different value of width in 90nm

From above discussion it is concluded that as width increases, speed get increased because propagation delay increases as capacitor takes more time to charge and discharge, but area improves.

3.3 Load capacitance Effect:

Capacitor mostly used to reduce any ripples coming in the power supply voltages [8],[9]. Basically it is used for elimination of offset voltages, so capacitor value should provide low to reduce the error in the output voltages. Smaller is the Capacitor, faster is the output value. Capacitor equation is shown below:

T=RC Equation (1)

From the fig.11 is shown that the C=0.1pF increases to 2pF then delay also increases from 1 to 4 lines and design becoming slower. Same result for fig.12.

180nm technology

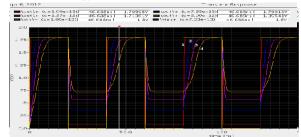


Fig.11: Simulation result for Different value of load verses Vin in 180nm

90nm technology

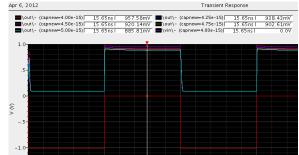


Fig.12: Simulation result for Different value of load verses Vin in 90nm

In this paper load capacitance 100fF is used for 180nm and 4fF is used for 90nm which is connected at the Vout+ and Vout-, it concluded that more capacitor causes more delay.

4 ANALYSIS OF DELAY

Propagation delay is the time taking by the input to change to produce corresponding change in the output signal. Design is faster as the delay reduces, in the circuit.

Below are given the transient response of the design in both 180nm and 90nm technology.[10]

180nm technology

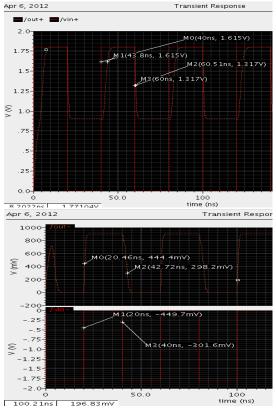


Fig.13: Propagation delay of the 180nm

90nm technology

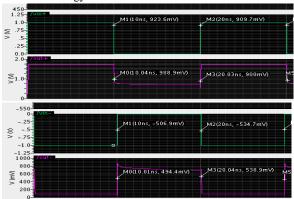


Fig.14: Propagation delay of the 90nm

Table.3 COMPARISONS OF THE PROPAGATION DELAYS

		Tplh	Tphl
180nm	Vout+	3.8ns	0.51ns
	Vout-	0.46ns	2.72ns
90nm	Vout+	0.03ns	0.04ns
	Vout-	0.01ns	0.04ns

So form the above Table.3 it is shown that for both the output delay is less in 90nm compares to 180nm technology.

5 POWER ANALYSIS

Power dissipation is the rate at which energy is dissipated from electrical circuits. It is measured in watts.

P=Idd.vdd Equation (2)

For the comparator design power dissipation increases as area increases due to leakage form circuits, so in 90nm it is concluded from above discussion that area is smaller than the 180nm so power dissipation is reduced in 90nm process technology and it get more power saving design.

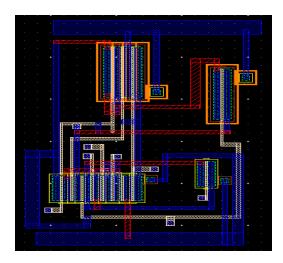
6 RESULT TABLE

Table.4 Comparison Result of Comparator Design

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	180nm design	Present			
		Design(90nm)			
length	180nm	90nm			
Load	100fF	4fF			
capacitance					
Clock	50Mhz	50Mhz			
frequency					
Wp/Wn	6µm/3µm	0.24 μm/0.12 μm			
Voltage Supply	1.8V	1V			
Delay	1.59ns	0.035ns			
a	44.5.337	20.24 11			
Static power	41.7pW	20.34pW			
dissipation					
A	172 15W	60.91W			
Average power	173.15μW	69.81 μW			
dissipation					
Equivalent	9.97µV/sq(Hz)	8.45µV/sq(Hz)			
output noise	<i>λ.λ γ</i> μ ν/δη(112)	σ. 4 5μ γ/sq(112)			
output noise					

7 LAYOUT

Layout of presently designed Comparator is shown below in 180nm process technology and its DRC and LVS report is also checked.



8. CONCLUSION

In this paper the CMOS Charge Sharing Latch Comparator is implemented in both 90nm and 180nm technology. Its simulation results and various parameters is compared for both technology. Here the input voltages used are 0.9V and 1.8V.The implemented new design shows reduction of power and enhancement of speed in lower technology than 180nm, comparision result is shown in Table.3.

From the result, it is found that the average power dissipation is reduced approximately 61% in 90nm compared to the 180nm and the delay minimizes to 0.035ns. Hence it is concluded that in 90nm Comparator operate with higher speed and provide more stabilized Output compare to 180nm.

9. REFERENCES

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