Performance Improvement of 4x4 Extended Alamouti Scheme with Implementation of Eigen Beamforming Technique

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ABSTRACT

This paper highlights on the error performance improvement of 4 ×4 Extended Alamouti Scheme with implementation of Eigen-Beamforming technique. In a MIMO communication system, the beamforming technique improves the Signal-to-Noise Ratio (SNR) and/or Signal-to-Interference-plus-Noise ratio. So, to get advantage of beamforming technique, 4x4 Extended Alamouti Scheme is implemented with Eigen beamforming technique. This combined technique comprises of channel state information which is fed back to the transmitter side in order to achieve proper error performance as compared to Extended Alamouti Scheme without Beamforming. This proposed scheme provides array gain by increasing the bit-error-rate (BER) performance for different wireless communication channels and M-PSK modulation scheme.

Keywords

MIMO, BER, SVD, EAC, PSK

1. INTRODUCTION

If we want to build any digital communication system with high data rates, then we need to increase the order of the modulation scheme, because the data rate of the QPSK modulation scheme is higher than that of the BPSK modulation scheme. At the same time as we move towards the higher order modulation scheme, along with increased in data rates, error rates are also increased. So, if we want to achieve small error rates than we need to implement MIMO transmission systems [9].

Mainly three types of antenna configurations available those are SIMO (Single Input Multiple Output), MISO (Multiple Input Single Output) and MIMO (Multiple Input Multiple Output).

When there is only one transmitting antenna and more than one receiving antennas are available in any communication system than that system is known as Single Input Multiple Output (SIMO) system, shown in figure-1(a). If a communication system consisting of more than one transmitting antennas and one receiving antennas than that system is known as Multiple Input Single Output (MISO) communication system, shown in figure-1(b). Similarly if a Nishchal M. Rindani Sr. Lecturer, EC Department A.V.P.T.I., Rajkot, India- 360001

communication system consisting of more than one transmitting as well as receiving antennas than that system is known as MIMO (Multiple Input Multiple Output), shown in figure-1(c). [11], [12], [16]

Now, if we want to use these multiple antenna configurations than there should be proper transmission schemes. To decrease the error rate and to increase the data rate and signal to noise ratio are available three transmission schemes: Spatial Multiplexing, Diversity and Beamforming, see figure-2 [5].



Fig 1: Different Antenna Configurations

By using SPATIAL MULTIPLEXING MIMO transmission scheme, we get benefit in improvement in data rate. That is achieved by improvement in MULTIPLEXING GAIN. By use of BEAMFORMING, we get improvement in SNR and that achieve by increment in ANTENNA GAIN. DIVERSITY technique used to decrease the error rate by increasing CODING GAIN as well as DIVERSTIY GAIN.

There is a tradeoff between all these gains. So, if we want to overcome this tradeoff, then the combined technologies are useful [1].



Fig 2: MIMO Transmission Scheme

2. EXTENDED ALAMOUTI CODING

As discussed in [7], we can improve the error performance by transmitting same symbols for more than one time by using the concepts of coding and interleaving. In [7], one of the well known space time transmitting for 2x1 diversity order as well as hybrid for 2x2 diversity order are described. Now, in [15], basic idea of extension in Alamouti scheme provided, which has obtained by extending it for 4x1-diversity order. Now, here it further extended for 4x4-diversity order [17].Now, consider block diagram representation of EAC (Extended Alamouti Code) shown in figure-3.

"Firstly, four symbols S1, S2, S3 & S4 are given to the STBC encoder, which uses the concept of coding and interleaving. The symbols are transmitted by different antennas through different paths of the same wireless channel.. Similarly during second time-slot its coded symbols such as S2*, -S1*, S4* and -S3* are being transmitted and so on.



Fig 3: Extended Alamouti Scheme

All transmitted coded symbols passed through the wireless communication channel and this channel is generally represented by channel matrix shown in equation-(1):-

$$\mathbf{H} = \begin{bmatrix} \mathbf{h}_{1,1} & \mathbf{h}_{1,2} & \mathbf{h}_{1,3} & \mathbf{h}_{1,4} \\ \mathbf{h}_{2,1} & \mathbf{h}_{2,2} & \mathbf{h}_{2,3} & \mathbf{h}_{2,4} \\ \mathbf{h}_{3,1} & \mathbf{h}_{3,2} & \mathbf{h}_{3,3} & \mathbf{h}_{3,4} \\ \mathbf{h}_{4,1} & \mathbf{h}_{4,2} & \mathbf{h}_{4,3} & \mathbf{h}_{4,4} \end{bmatrix}$$
(1)

where hi,j represents the fading coefficient of the propagation path from the ith transmit antenna to the jth receive antenna, with i=1..NT, j=1..NR.

These all transmitted symbols are being received at EAC decoder / EAC decoder, which are given by the following well-known equation:-

$$R = H^*S + \eta \tag{2}$$

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Where S represents the transmitting symbol matrix as shown in figure, H represents the channel matrix as shown in equation-(1) and η is AWGN (Additive White Gaussian Noise) [16]. All these received symbols are given to the channel estimator and combiner section of decoder portion of MIMO communication system. These section estimates the channel parameter from the received symbols and given to the combiner section. Now outputs of combiner and channel estimator portion are given by –

$$\hat{S} = H^{\dagger} * Y$$
 (3)

Where Y and H+ are represented by equation(4) and (5), respectively. Now, this \hat{s} represents the symbols detected at the receiver side, which are being compared with the transmitted symbols S.

$$\mathbf{H}^{\dagger} = \begin{pmatrix} \mathbf{h}_{1,1} & \mathbf{h}_{1,2} & \mathbf{h}_{1,3} & \mathbf{h}_{1,4} \\ \mathbf{h}_{2,1} & \mathbf{h}_{2,2} & \mathbf{h}_{2,3} & \mathbf{h}_{2,4} \\ \mathbf{h}_{3,1} & \mathbf{h}_{3,2} & \mathbf{h}_{3,3} & \mathbf{h}_{3,4} \\ \mathbf{h}_{4,1} & \mathbf{h}_{4,2} & \mathbf{h}_{4,3} & \mathbf{h}_{4,4} \\ -\mathbf{h}_{2,1}^{*} & \mathbf{h}_{1,1}^{*} & -\mathbf{h}_{2,4}^{*} & \mathbf{h}_{2,3}^{*} \\ -\mathbf{h}_{2,2}^{*} & \mathbf{h}_{2,1}^{*} & -\mathbf{h}_{2,4}^{*} & \mathbf{h}_{2,3}^{*} \\ -\mathbf{h}_{3,2}^{*} & \mathbf{h}_{3,1}^{*} & -\mathbf{h}_{3,4}^{*} & \mathbf{h}_{3,3}^{*} \\ -\mathbf{h}_{4,2}^{*} & \mathbf{h}_{4,1}^{*} & -\mathbf{h}_{4,4}^{*} & \mathbf{h}_{4,3}^{*} \\ -\mathbf{h}_{2,3}^{*} & -\mathbf{h}_{2,4}^{*} & \mathbf{h}_{2,1}^{*} & \mathbf{h}_{4,2}^{*} \\ -\mathbf{h}_{3,3}^{*} & -\mathbf{h}_{2,4}^{*} & \mathbf{h}_{2,1}^{*} & \mathbf{h}_{2,2}^{*} \\ -\mathbf{h}_{3,3}^{*} & -\mathbf{h}_{2,4}^{*} & \mathbf{h}_{2,1}^{*} & \mathbf{h}_{3,2}^{*} \\ -\mathbf{h}_{3,3}^{*} & -\mathbf{h}_{4,4}^{*} & \mathbf{h}_{4,1}^{*} & \mathbf{h}_{4,2}^{*} \\ -\mathbf{h}_{4,3}^{*} & -\mathbf{h}_{4,4}^{*} & \mathbf{h}_{3,1}^{*} & \mathbf{h}_{3,2}^{*} \\ -\mathbf{h}_{4,3}^{*} & -\mathbf{h}_{4,4}^{*} & \mathbf{h}_{4,1}^{*} & \mathbf{h}_{4,2}^{*} \\ \mathbf{h}_{1,4} & \mathbf{h}_{1,3} & \mathbf{h}_{1,2} & \mathbf{h}_{1,1} \\ \mathbf{h}_{2,4} & \mathbf{h}_{2,3} & \mathbf{h}_{2,2} & \mathbf{h}_{2,1} \\ \mathbf{h}_{3,4} & \mathbf{h}_{3,3} & \mathbf{h}_{3,2} & \mathbf{h}_{3,1} \\ \mathbf{h}_{4,4} & \mathbf{h}_{4,3} & \mathbf{h}_{4,2} & \mathbf{h}_{4,1} \end{pmatrix}$$

(4)

$$y = H^* W^* \Theta^* S + \eta$$
⁽⁷⁾

(5) Where, y is the received signal, H shown in equation-(1). S is the t given by equation- (8). η is A

3. EXTENDED ALAMOUTI CODING WITH BEAMFORMING TECHNIQUE

 R^{1} R^{2*}

R^{3*}

Beamforming is one of the well-known MIMO transmission techniques implemented in order to improve the error performance by improving Signal-to-Interference-Noise Ratio (SINR) or Signal-to-Noise Ratio (SNR).

In case of the MIMO beamforming technique, we are providing information regarding angles at which they received at the antennas. It is necessary to provide these information, because signals are coming at the receiver through multipath fading effect which causes the signals to come at different time and angles. In addition, there is necessary to provide beamforming-weighting factors in order to provide proper shape to RF antennas. [9], [2].

Now, this angle information is developed by taking angle of the exponential of the channel fading parameter [18]. Similarly, weighting factor can be obtained by Eigen beamforming method.

Now, this Eigen beamforming method is based on SVD (Singular Value Decomposition) method. If we generate SVD of channel matrix H (shown in equation- (1)), than it results into following equation-(6);

$$SVD{[H]} = U*D*U^{T}$$
(6)

Where, D represents the diagonal matrix that represents the system as equivalent parallel SISO channel.U is the nonorthogonal matrices, which play important roles in the formation of the beamforming weights. [4]

Now, implementation and combination of MIMO Eigen Beamforming technique with EAC is shown in figure – (4). Its Mathematical description is also shown below-



Fig 4: EAC with Beamforming Technique

Where, y is the received signal, H is the channel matrix as shown in equation-(1). S is the transmitted symbol matrix given by equation- (8). η is AWGN added during the transmission of the symbols through fading channel.

$$S = \begin{pmatrix} s_1 & s_2^* & s_3^* & s_4 \\ s_2 & -s_1^* & s_4^* & -s_3 \\ s_3 & s_4^* & -s_1^* & -s_2 \\ s_4 & -s_3^* & -s_2^* & s_1 \end{pmatrix}$$
(8)

Here, W is the beamforming weighting matrix, which is equal to the matrix U which represents Eigen vectors. This provides the channel state information back to the transmitter side.

Now, Θ is the angle information that is the inverse of the phase of the channel that is multiplied with weighting factor to ensure that the signals add constructively at the receivers [18]. That angle information in mathematical form is shown in below equation – (9):

$$\Theta = \begin{pmatrix} e^{-i\theta_{11}} & e^{-i\theta_{21}} & e^{-i\theta_{31}} & e^{-i\theta_{41}} \\ e^{-i\theta_{12}} & e^{-i\theta_{22}} & e^{-i\theta_{32}} & e^{-i\theta_{42}} \\ e^{-i\theta_{13}} & e^{-i\theta_{23}} & e^{-i\theta_{33}} & e^{-i\theta_{43}} \\ e^{-i\theta_{14}} & e^{-i\theta_{24}} & e^{-i\theta_{34}} & e^{-i\theta_{44}} \end{pmatrix}$$
(9)

Now, these received signals are given to the decoder and at last to the ML detector. To measure the error performances for various channels such as AWGN channel and RAYLEIGH and NAKAGAMI fading channels, the detected symbols are compared with the transmitted symbols.

4. SIMULATION RESULTS & DISCUSSION

This section comprises of the simulation results of the ALAMOUTI 2x2 scheme as shown in [1] with beamforming, EAC with and without beamforming technique for M-PSK modulation schemes for various wireless fading channels.

Here, figure-(5), (6) shows the results for Rayleigh fading channels and M-PSK modulation schemes. As shown in these figures, Extended Alamouti Code for 4x4-diversity order with beamforming provides better results as compared to the Alamouti 2x2 with beamforming technique.

Along with that, it can be seen from the figures EAC with beamforming provides better coding as well as diversity gain as compared to the EAC without beamforming technique. Therefore, we can say that as the fading effect is higher, this combining technique provides better error performance because of CSI (Channel State Information) feed back to the transmitter side.



Fig 5: BER analysis for Rayleigh fading channel, BPSK Modulation Scheme



FIg 6: BER analysis for Rayleigh fading channel, QPSK Modulation Scheme

Similarly, as the modulation scheme order is increased from '2' to '4', data rate is increased. Therefore, along with it error rate is also increased. Therefore, error performance is degraded.



Fig 7: BER analysis for NAKAGAMI fading channel, BPSK Modulation Scheme (EAC)



Fig 8: BER analysis for NAKAGAMI fading channel, BPSK Modulation Scheme (Alamouti 2x2 with BF)

Now, figures 7 to 12 shows the error performances for the NAKAGAMI fading channel for M-PSK modulation scheme for Extended Alamouti Scheme, EAC 4x4 with Beamforming and Alamouti scheme with beamforming, respectively. Now, as the value of nakagami factor is increasing from m=0.5 to m=2, its error performance is improved for each and every modulation schemes and diversity techniques. Now, if we compare EAC with and without Beamforming techniques for M-PSK modulation schemes than EAC with beamforming provides better results as compared to the EAC without beamforming as shown in Table-1.

Table-1. BER ANALYSIS FOR QPSK

\searrow	Nakagami Fading Channel	BER Values	
SNR		EAC 4x4	EAC 4X4 + BF
At SNR=5dB	m=0.5	0.22885	0.1302
	m=2	0.059231	0.0186
At SNR=10dB	m=0.5	0.1359	0.0563
	m=2	0.010577	0.0077



Fig 9: BER analysis for NAKAGAMI fading channel, BPSK Modulation Scheme (EAC with BF)



Fig 10: BER analysis for NAKAGAMI fading channel, QPSK Modulation Scheme (EAC)



Fig 11: BER analysis for NAKAGAMI fading channel, QPSK Modulation Scheme (Alamouti 2x2 with BF)



Fig 12: BER analysis for NAKAGAMI fading channel, QPSK Modulation Scheme (EAC with BF)



Fig 13: BER analysis for AWGN channel, BPSK Modulation Scheme



Fig 14: BER analysis for AWGN channel, QPSK Modulation Scheme

Now, figure-13 and 14 provide the results for the all proposed schemes for AWGN channel and M-PSK modulation schemes. Here also proposed scheme i.e. EAC with Beamforming provides the better error performance as compared to other diversity schemes.

5. CONCLUSION

This Paper provides new proposed scheme, which is combination of Extended Alamouti Code and Eigen-Beamforming technique for various wireless communication channels such as AWGN, RAYLEIGH & NAKAGAMI fading channels. As shown in section of result and analysis, this new proposed combined scheme of EAC 4x4 with beamforming provides the better error performance as compared to others. Here, error performance is improved not only in terms of coding gain but also in terms of diversity gain.

6. REFERENCES

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